

[0016] FIG. 1 is a simplified schematic diagram of circuit 20 wherein ESD clamp 21 is placed, for example, between input-output (I/O) terminal 22 and ground or common terminal 23 of an IC to protect other devices therein, that is, to protect “circuit core” 24 also coupled to I/O and common terminals 22, 23. Person of skill in the art will understand that ESD clamp 21 may be placed across any terminals of the IC, and reference herein to I/O terminals is intended to include any and all other terminals not merely those used for input or output signals. Further, the Zener diode symbol illustrated in block 21 of FIG. 1 is merely for convenience of identifying the voltage limiting function of ESD block 21 and not intended to imply that a Zener diode is necessarily present therein. With respect to structures or elements used for ESD protection, the terms device, clamp and transistor are used interchangeably.

[0017] FIG. 2 is a simplified schematic diagram illustrating internal components of ESD clamp 21 employing bipolar transistor 25, having emitter 26, collector 27, base 28, and internal resistance 29, coupled across I/O terminals 22, 23. When the voltage across terminals 22, 23 rises beyond a predetermined limit, bipolar transistor 25 turns on, limiting the voltage across terminals 22, 23, desirably to a level below that capable of damaging circuit core 24.

[0018] FIG. 3 shows simplified plot 30 of transmission line pulse current (I) versus voltage (V) for a typical electrostatic discharge (ESD) protection device such as, for example, device 21 of FIG. 2. As the applied voltage is increased, very little current flows until triggering voltage 31 is reached at voltage V_{t1} . Once triggered into operation, the ESD device conducts and the current increases to holding point 32 with current I_h and voltage V_h . Depending upon the internal impedance of the voltage source, current and voltage may further increase to point 33 at current I_{t2} and voltage V_{t2} , beyond which destructive failure may occur leading to further current increase accompanied by voltage decrease.

[0019] Electrostatic discharge (ESD) protection devices are intended to remain quiescent during normal operation of the associated semiconductor (SC) device(s) or non-SC device(s) or integrated circuit (IC) (i.e., the protected element (s) or circuit core 24) having a normal operating voltage V_o , but turn on when excessive voltage arises, thereby preventing damage to the protected element(s). The triggering voltage V_{t1} of the ESD device should exceed the maximum normal DC operating voltage $V_o(\text{MAX})$ of the protected elements, otherwise the ESD device will interfere with normal operation of the protected elements. Further, V_{t1} should be less than, for example, a voltage V_{TR} (usually a transient voltage) large enough to damage the protected element(s), hereafter referred to as the protected element break-down voltage, abbreviated as $V_{TR}(\text{PEBD})$. Thus, the ESD device should be designed so that $V_o(\text{MAX}) < V_{t1} < V_{TR}(\text{PEBD})$. It is commonplace to provide a number of ESD clamps 21 on a SC die or wafer so that all of the various terminals of the IC or electronic assembly may be protected. It is usually important that the V_{t1} values for these several ESD clamps be similar or lie within a narrow predetermined range.

[0020] When bipolar transistor 25 of FIG. 2 is used for ESD clamp 21, the triggering voltage V_{t1} is heavily influenced by the base-collector spacing of transistor 25. Unfortunately, it is often the case that even though several clamp transistors 25 are manufactured at the same time on the same SC wafer or die using nominally the same mask shapes and dimensions, there can be significant variation in the base-collector spacing

in different parts of the SC wafer and/or die as a function, for example, of the azimuthal orientation of transistor 25 on the wafer or die. This has the result that V_{t1} of nominally identical clamp devices can be different in different regions of the same IC, depending, for example, on their relative azimuthal orientation on the IC wafer or die. This V_{t1} variation (hereafter ΔV_{t1}) can adversely affect overall manufacturing yield and is not desirable. In the past, various process modifications have been used to minimize ΔV_{t1} , but such modifications are often accompanied by an undesirable increase in manufacturing cost or other difficulties.

[0021] Accordingly, there is an ongoing need to provide improved ESD clamps that operate at more consistent triggering voltages V_{t1} independent of their location or orientation on a particular IC die or wafer, that is, to minimize ΔV_{t1} . Further, it is desirable that the improved ESD clamps be obtainable without significant modification of the manufacturing process used for forming the ESD clamps and the associated circuit core of the IC. Furthermore, other desirable features and characteristics of the present invention will become apparent from the subsequent detailed description of the invention and the appended claims, taken in conjunction with the accompanying drawings and this background of the invention.

[0022] FIG. 4 shows a simplified cross-sectional view of ESD clamp transistor 70 implemented in a semiconductor substrate according to an embodiment of the present invention. Transistor 70 is formed in substrate 72 (e.g. P) having buried layer region 73 therein (e.g., N type, abbreviated as NBL 73). Overlying NBL 73 is region 74 extending from NBL 73 to upper surface 71. Region 74 may be either N or P, depending upon the particular process available and the desired devices being fabricated at the same time. Hence, region 74 is identified in the various drawings as being “N/P” indicating that either conductivity type may be used. In a preferred embodiment, P type is used. Within region 74 are shallow trench dielectric isolation (STI) regions 79 having interface 791 with the underlying semiconductor, WELL regions 761 and 762 (e.g., N type, collectively 76) with contact region 80 (e.g., N+) in region 762, and WELL region 75 (e.g., P type) with lateral interface or boundary 751. Further N region 86 is provided in electrical contact with N WELL region 762 with lateral interface or boundary 861 facing toward interface 751 of P WELL region 75 and separated therefrom by distance D. Doped contact region 77 (e.g., P+) is provided in P WELL region 75 to make Ohmic contact to P WELL region 75. Doped region 78 (e.g., N+) in P WELL region 75 serves as the emitter of transistor 70. P WELL region 75 serves as the base of transistor 70. N WELL regions 86, 762 with N+ contact region 80 serve as the collector of transistor 70. Intermediate portion 85 of region 74 lies between interfaces or boundaries 751, 861 separated by distance D. As will be subsequently explained, intermediate portion 85 of region 74 is more lightly doped than adjacent regions 75 and 86. As the voltage applied between terminals 22, 23 of transistor 70 approaches avalanche conditions, more lightly doped intermediate portion 85, whether of N or P type, becomes substantially depleted of free carriers, and effectively defines the base-collector spacing. Hence, for convenience of explanation, distance D across intermediate portion 85 between boundaries 751, 861 is referred to herein as the base-collector spacing irrespective of whether portion 85 is N type or P type. Dielectric layer 81 is conveniently provided on surface (e.g., interface) 71 with openings therein extending to